

EVALUATING "A+B = K" CONDITIONS IN CONSTANT TIME

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ABSTRACT

One of the most important components of an ALU is the adder. Its response time is mainly determined by the carry propagation delay. Evaluation of conditions between two numbers are usually performed with the ALU by means of a subtraction. In this paper we deal with a type of conditions that can be evaluated without requiring a complete ALU operation. The circuit that is presented detects the condition $A+B=K$ (n-bit numbers) in constant time, avoiding the carry propagation delay. Some applications for this circuit are also presented.

INTRODUCTION

The carry computation is one major problem in the response time of parallel adders. Several approaches have been proposed in order to reduce it [1][2][3]. VLSI techniques have been also used to minimize design costs and chip area. The fastest adders, such as lookahead adders, perform additions of n-bit numbers in time $O(\log n)$ and area $O(n \log n)$.

This paper deals with a problem associated with parallel adders. A circuit for detecting when the addition of two n-bit numbers is equal to another n-bit number ($A+B=K$) is presented. We will prove that the result of this evaluation can be computed in constant time and area $O(n)$, avoiding the problem of the carry propagation delay.

The paper is organized as follows. First, the theoretical basis of the problem is presented. Next, the design of the circuit is described. Finally, some applications for this circuit are discussed.

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THEORETICAL BASIS

Given three n-bit vectors $A=a_n a_{n-1} \dots a_1$, $B=b_n b_{n-1} \dots b_1$ and $K=k_n k_{n-1} \dots k_1$ that represent two's complement integers, we want to design a circuit that can evaluate the condition $A+B=K$ (arithmetic addition) in constant time. This means that the evaluation does not depend on the length (number of bits) of the vectors.

The basic idea of the circuit is the local evaluation of the condition at each bit position i , assuming that the condition is fulfilled in the rest of bits. The last stage computes the global logical OR of all the local evaluations.

The behavior of a full adder can be described with the following expressions:

$$\begin{aligned} p_i &= a_i \oplus b_i && \text{(Carry propagation)} \\ g_i &= a_i \wedge b_i && \text{(Carry generation)} \\ c_i &= (p_i \wedge c_{i-1}) \vee g_i && \text{(Carry. We define } c_0=0) \\ r_i &= p_i \oplus c_{i-1} && \text{(Addition result)} \end{aligned}$$

The condition " $A+B=K$ " could be detected by comparing r_i and k_i at each bit position i and performing a global OR of all the comparisons. In this case the response time would be determined by the carry propagation time, since c_i is defined as a function of c_{i-1} . In order to avoid the carry propagation problem, we define the following expressions:

$$\begin{aligned} q_i &= (p_i \wedge \bar{k}_i) \vee g_i && \text{(Predicted carry. We define } q_0=0) \\ s_i &= p_i \oplus q_{i-1} && \text{(Predicted addition result)} \\ z_i &= s_i \oplus k_i && (z_i = 0 \Leftrightarrow s_i = k_i) \\ Z_k &= \bigvee_{i=1}^k z_i \end{aligned}$$

The predicted carry (q_i) substitutes the carry for the computation of the predicted result. We can easily prove that $q_i = c_i$ in case that $r_i = k_i$:

$$q_i = (p_i \wedge \bar{k}_i) \vee g_i = (p_i \wedge \bar{r}_i) \vee g_i = (p_i \wedge c_{i-1}) \vee g_i = c_i$$

The design of the circuit is based on the following theorem:

Theorem.

$$Z_n = 0 \Leftrightarrow r_i = k_i \text{ for each } i \in \{1, \dots, n\}$$

Proof. By induction on n .

First, we will prove that the theorem holds for $n=1$.

$$Z_1 = z_1 = s_1 \oplus k_1 = p_1 \oplus k_1 = r_1 \oplus k_1 \quad (\text{since } c_0 = q_0 = 0)$$

thus

$$Z_1 = 0 \Leftrightarrow r_1 = k_1$$

Next, we will prove that the theorem holds for $n > 1$. Let's assume the theorem holds for $n-1$ bits.

If $Z_{n-1} = 1$ then $Z_n = Z_{n-1} \vee z_n = 1$, and the theorem holds for n bits, since there is $i \in \{1, \dots, n-1\} \subset \{1, \dots, n\}$ such that $r_i \neq k_i$ (by induction hypothesis).

If $Z_{n-1} = 0$ then $Z_n = z_n$

By the definition of z_i we have that

$$z_n = s_n \oplus k_n = p_n \oplus q_{n-1} \oplus k_n$$

since $Z_{n-1} = 0$ then $r_{n-1} = k_{n-1}$ and $q_{n-1} = c_{n-1}$ (by induction hypothesis). Therefore,

$$z_n = p_n \oplus c_{n-1} \oplus k_n = r_n \oplus k_n$$

and

$$z_n = 0 \Leftrightarrow r_n = k_n$$

Since $r_i = k_i$ for each $i \in \{1, \dots, n-1\}$, we have that

$$Z_n = z_n = 0 \Leftrightarrow r_n = k_n \Leftrightarrow r_i = k_i \text{ for each } i \in \{1, \dots, n\}$$

□

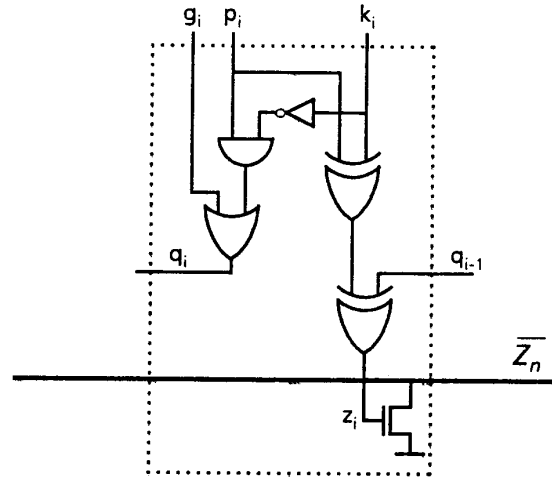


Figure 1. Circuit design

CIRCUIT DESIGN

From the previous section we can observe that function q_i (predicted carry) does not depend on any information in the other stages. So functions z_i can be computed in parallel. Figure 1 depicts a diagram of the circuit that computes Z_n . We assume that each gate and the \bar{Z}_n line discharge (n -input NOR) take constant time. So we can state that the computation of Z_n also takes constant time. Since the circuit is made of n identical cells, the chip area is $O(n)$.

The circuit design can be simplified if any of the operands is constant. Figure 2 shows a diagram of the circuit when vector K is constant.

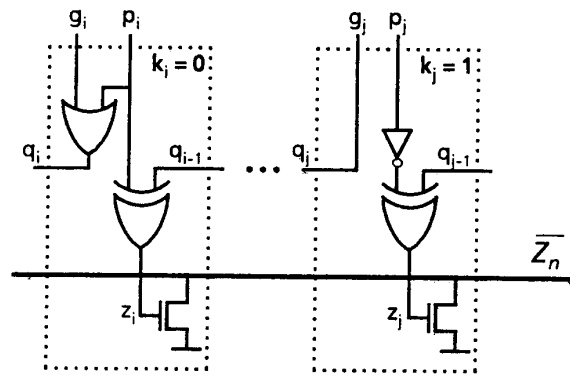


Figure 2. Circuit design with K constant

SOME APPLICATIONS

The condition evaluation is one of the most important operations performed in the execution of conditional branches. In most architectures, the condition is evaluated as a function of the condition codes. Their value depends on the result of ALU operations. Figure 3 depicts a widely used ALU structure [4]. It consists of an Operand Modifier Unit (OMU) and an adder. The OMU computes the carry propagation and generation functions (p_i and g_i) depending on the data input (a_i and b_i) and the operation.

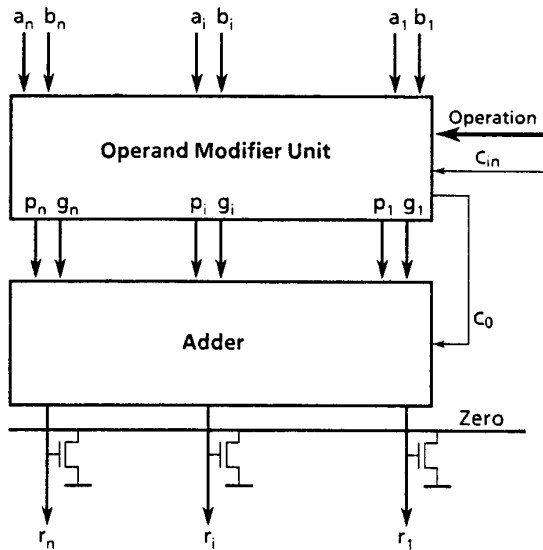


Figure 3 ALU structure

The dependencies produced by the execution of branches have been extensively studied by many authors [5][6]. In pipelined processors, the condition evaluation has to be delayed until all the instructions that modify the condition codes and precede the branch have finished its execution. The condition evaluation is the most important dependency that restricts the execution of branches with zero delay [7].

Katevenis observed that 80% of conditional branches involve tests for equality, inequality and any relation with zero (fast comparisons) [8]. Equality and inequality tests are determined by the zero condition code (Z), usually after a subtraction operation (comparison). Relations with zero are determined by Z and the sign bit of the operand that is tested.

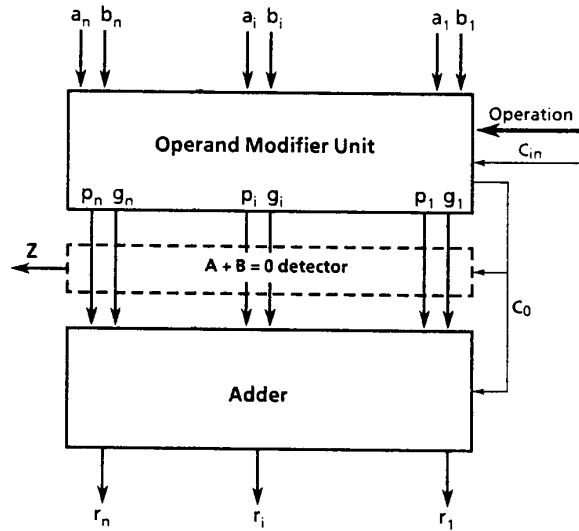


Figure 4 Evaluation of condition code Z

By designing a circuit that detects the condition $A+B=0$, the computation of Z can be advanced and the delay produced by the condition evaluation reduced (see figure 4). This improvement can increase the processor performance substantially since about one in every six instructions are branches that have to evaluate this kind of conditions.

Another application can be found for DO-like loops. This kind of loops are very used in numerical programs (figure 5.a). The compiler generates a code similar to the one shown in figure 5.b. As Katevenis also observed, some comparisons can be converted to fast comparisons [6]. This is the case of " $I \leq N$ ", which can be converted to " $I \neq N+1$ " without modifying the loop behavior. By considering an ALU structure such as the one shown in figure 6, and introducing a new instruction in the machine language (NEXT), the compiler can generate a code similar to the one in fig. 5.c. The

do $i=1,n$	$R_i \leftarrow 1$	$R_i \leftarrow 1$
.	do:	$R_{limit} \leftarrow n+1$
.	.	do:
.	.	.
end do	$R_i \leftarrow R_i + 1$.
	if $R_i \leq n$ goto do	NEXT R_i, do
(a)	(b)	(c)

Figure 5. Do-like loops

instruction NEXT increments R_i and compares the new value with R_{limit} . In case that $R_i \neq R_{limit}$, control is transferred to the branch target address. Again, the improvement is based on the fact that the condition can be evaluated before computing the new value of R_i , avoiding pipeline delays in the execution of conditional branches.

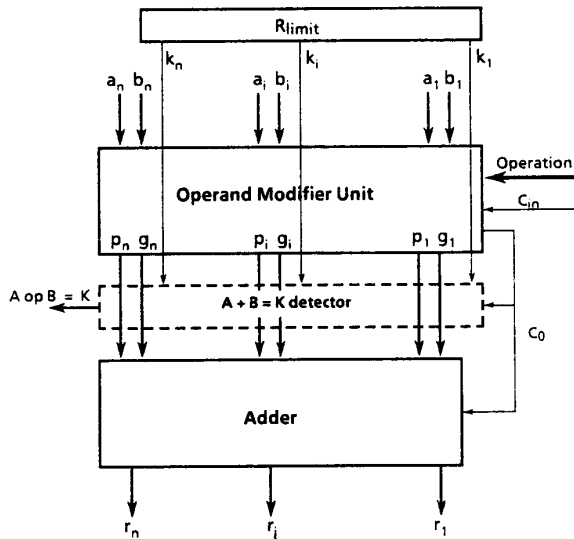


Figure 6. ALU structure for DO-like loops.

CONCLUSIONS

In this paper we have presented a circuit that detects the condition " $A+B=K$ " in constant time. Its area is proportional to the number of bits of the vectors. The theoretical basis and several design approaches have been described.

This circuit can be used to detect a wide spectrum of conditions in branch instructions. It can improve the processor performance by advancing the evaluation of conditions and eliminating the pipeline delays produced by these operations.

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